



# Introduction to Space-Time Coding

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## Last Time

1. A fully diverse space-time code is a family  $\mathcal{C}$  of (square) complex matrices such that  $\det(\mathbf{X} - \mathbf{X}') \neq 0$  when  $\mathbf{X} \neq \mathbf{X}'$ .
2. Division algebras whose elements can be represented as matrices satisfy full diversity by definition.
3. Hamilton's quaternions provide such a family of fully diverse space-time codes.

# Outline

## Division Algebras

- Cyclic Algebras

- Crossed Product Algebras

## Quotients of Space-Time Codes

- $2 \times 2$  Space-Time Coded Modulation



## Cyclic Algebras: Definition

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- *Multiplication* rule:  $\lambda e = e\sigma(\lambda)$ ,  $\sigma : L \rightarrow L$ , the generator of the Galois group of  $L/\mathbb{Q}(i)$ .



## Cyclic Algebras: Coding ( $n = 2$ )

- For  $n = 2$ , compute the *multiplication* by  $x$  of  $y \in \mathcal{A}$ :

$$\begin{aligned}
 xy &= (x_0 + ex_1)(y_0 + ey_1) \\
 &= x_0y_0 + e\sigma(x_0)y_1 + ex_1y_0 + \gamma\sigma(x_1)y_1 && \lambda e = e\sigma(\lambda) \\
 &= [x_0y_0 + \gamma\sigma(x_1)y_1] + e[\sigma(x_0)y_1 + x_1y_0] && e^2 = \gamma
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2. In the basis  $\{1, e\}$ , we have

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3. Correspondence between  $x$  and its *multiplication matrix*.

$$x = x_0 + ex_1 \in \mathcal{A} \leftrightarrow \begin{pmatrix} x_0 & \gamma\sigma(x_1) \\ x_1 & \sigma(x_0) \end{pmatrix}.$$

# Cyclic Algebras: Encoding

- In general:

$$x \leftrightarrow \begin{pmatrix} x_0 & \gamma\sigma(x_{n-1}) & \gamma\sigma^2(x_{n-2}) & \dots & \gamma\sigma^{n-1}(x_1) \\ x_1 & \sigma(x_0) & \gamma\sigma^2(x_{n-1}) & \dots & \gamma\sigma^{n-1}(x_2) \\ \vdots & & \vdots & & \vdots \\ x_{n-2} & \sigma(x_{n-3}) & \sigma^2(x_{n-4}) & \dots & \gamma\sigma^{n-1}(x_{n-1}) \\ x_{n-1} & \sigma(x_{n-2}) & \sigma^2(x_{n-3}) & \dots & \sigma^{n-1}(x_0) \end{pmatrix}.$$

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- Every  $x_i \in L$  *encodes*  $n$  information symbols.

# Cyclic Division Algebras

- *Remember*: Given  $L/\mathbb{Q}(i)$ , a cyclic algebra  $\mathcal{A}$  is defined by

$$\mathcal{A} = \{(x_0, x_1, \dots, x_{n-1}) \mid x_i \in L\}$$

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- **Proposition.** If  $\gamma$  and its powers  $\gamma^2, \dots, \gamma^{n-1}$  are not algebraic norms (there is no  $x \in L$  with  $N_{L/\mathbb{Q}(i)}(x) = \gamma^j$ ,  $j = 1, \dots, n-1$ ), then the cyclic algebra  $\mathcal{A}$  is a *division algebra*.

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To obtain *space-time codes*:

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To obtain *space-time codes*:

1. Take a *cyclic extension*  $L/\mathbb{Q}(i)$  of degree  $n$  (# antennas).
2. Build a *cyclic division algebra*.
3. This gives *fully diverse* codes and a practical encoding for *every*  $n$ .

[ F. Oggier, G. Rekaya, J.-C. Belfiore, E. Viterbo, "Perfect Space-Time Block Codes." ]

## An Example: the Golden Code

- The *Golden number* is  $\theta = \frac{1+\sqrt{5}}{2}$ , a root of  $x^2 - x - 1 = 0$  ( $\sigma(\theta) = \frac{1-\sqrt{5}}{2}$  is the other root).

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- Take  $L = \mathbb{Q}(i, \theta)$ , the cyclic extension  $L/\mathbb{Q}(i)$  and *the cyclic algebra which is division*

$$\mathcal{A} = \{y = (u + v\theta) + e(w + z\theta) \mid e^2 = i, u, v, w, z \in \mathbb{Q}(i)\}$$

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- We define the code  $\mathcal{C}$  by

$$\left\{ \begin{pmatrix} x_{11} & x_{12} \\ x_{21} & x_{22} \end{pmatrix} = \begin{pmatrix} a + b\theta & c + d\theta \\ i(c + d\sigma(\theta)) & a + b\sigma(\theta) \end{pmatrix} : a, b, c, d \in \mathbb{Z}[i] \right\}$$

## The Golden code: $\gamma = i$ not a norm (I)

- The determinant of  $\mathbf{X} \in \mathcal{C}$  is

$$\begin{aligned} \det(\mathbf{X}) &= \det \begin{pmatrix} a + b\theta & c + d\theta \\ i(c + d\sigma(\theta)) & a + b\sigma(\theta) \end{pmatrix} \\ &= (a + b\theta)(a + b\sigma(\theta)) - i(c + d\theta)(c + d\sigma(\theta)). \end{aligned}$$

- Thus

$$0 = \det(\mathbf{X}) \iff i = \frac{(a + b\theta)(a + b\sigma(\theta))}{(c + d\theta)(c + d\sigma(\theta))}$$

- Make sure  $\gamma = i$  is *not a norm*.

## The Golden code: $\gamma = i$ not a norm (II)

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- Consider

$$\mathbb{Q}_5 = \left\{ a_{-m} \frac{1}{5^m} + a_{-m+1} \frac{1}{5^{m-1}} + \dots + a_{-1} \frac{1}{5} + a_0 + a_1 5 + \dots \right\}$$

the field of 5-adic numbers, and

$$\mathbb{Z}_5 = \{a_0 + a_1 5 + a_2 5^2 + \dots\} = \{x \in \mathbb{Q}_5 \mid \nu_5(x) \geq 0\}$$

its valuation ring.



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its valuation ring.

- Then  $\mathbb{Q}(i)$  can be embedded into  $\mathbb{Q}_5$  by

$$i \mapsto 2 + 5\mathbb{Z}_5$$

(the polynomial  $X^2 + 1$  has roots in  $\mathbb{Z}_5$ , because it has roots in  $\mathbb{F}_5$ , then use Hensel's Lemma).

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- Let  $x = a + b\sqrt{5} \in K$  with  $a, b \in \mathbb{Q}(i)$  then we must show that

$$N_{L/\mathbb{Q}(i)}(x) = a^2 - 5b^2 = i$$

has no solution for  $a, b \in \mathbb{Q}(i)$ .

## The Golden code: $\gamma = i$ not a norm (III)

- We can lift the norm equation in the 5-adic field  $\mathbb{Q}_5$

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- We take the valuations of both sides:

$$\nu_5(a^2 - 5b^2) = \nu_5(2 + 5x)$$

to show that  $a$  and  $b$  must be in  $\mathbb{Z}_5$ .

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- Since  $x \in \mathbb{Z}_5$ ,  $\nu_5(2 + 5x) = \inf\{\nu_5(2), \nu_5(x) + 1\} = 0$ . Now,  $\nu_5(a^2 - 5b^2) = \inf\{2\nu_5(a), \nu_5(b) + 1\}$  must be 0, hence  $\nu_5(a) = 0$  which implies  $a \in \mathbb{Z}_5$  and consequently  $b \in \mathbb{Z}_5$ .

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- We conclude by showing that

$$a^2 - 5b^2 = 2 + 5x \quad a, b, x \in \mathbb{Z}_5$$

has no solution. Reducing modulo  $5\mathbb{Z}_5$  we find that 2 should be a square in  $\mathbb{F}_5$ , which is a contradiction.

## The Golden Code: Minimum Determinant

- Let  $\mathbf{X} \in \mathcal{C}$  be a codeword from the Golden code.

$$\begin{aligned}
 \det(\mathbf{X}) &= \det \begin{pmatrix} a + b\theta & c + d\theta \\ i(c + d\sigma(\theta)) & a + b\sigma(\theta) \end{pmatrix} \\
 &= (a + b\theta)(a + b\sigma(\theta)) - i(c + d\theta)(c + d\sigma(\theta)) \\
 &= a^2 + ab(\sigma(\theta) + \theta) - b^2 - i[c^2 + cd(\theta + \sigma(\theta)) - d^2] \\
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- Thus

$$\det(\mathbf{X}) \in \mathbb{Z}[i] \Rightarrow \delta_{\min}(\mathcal{C}) = |\det(\mathbf{X})|^2 \geq 1.$$

- Is a property of *rings of integers*, can be generalized in dimension  $n$ .

## The Golden code: a Space-Time lattice code (I)

- A complex lattice  $\Lambda$  is given by its *generator matrix*:

$$\Lambda = \{M\mathbf{v} \mid \mathbf{v} \in \mathbb{Z}[i]^n\}$$

- Note that  $\mathbf{X} \in \mathcal{C}$  can be written

$$\begin{aligned} \mathbf{X} &= \text{diag} \left( M \begin{bmatrix} a \\ b \end{bmatrix} \right) + \text{diag} \left( M \begin{bmatrix} c \\ d \end{bmatrix} \right) \cdot \begin{bmatrix} 0 & 1 \\ \gamma & 0 \end{bmatrix} \\ &= \begin{bmatrix} a + b\theta & c + d\theta \\ \gamma(c + d\sigma(\theta)) & a + b\sigma(\theta) \end{bmatrix}, \end{aligned}$$

where

$$M = \begin{bmatrix} 1 & \theta \\ 1 & \sigma(\theta) \end{bmatrix}.$$

- We add a structure of  $\mathbb{Z}[i]^2$  lattice on each layer to guarantee *no shaping loss*.

## The Golden code: a Space-Time lattice code (II)

- We recognize that

$$M = \begin{bmatrix} 1 & \theta \\ 1 & \bar{\theta} \end{bmatrix}$$

is the generator matrix of a lattice obtained from a quadratic number field.

- We add a structure of  $\mathbb{Z}[i]^2$  lattice on each layer by defining  $\mathcal{C}_{\mathcal{I}} \subset \mathcal{C}$  as

$$x_1, x_2, x_3, x_4 \in \mathcal{I} = (\alpha)\mathbb{Z}[i]\left[\frac{1+\sqrt{5}}{2}\right], \quad \alpha = 1 + i - i\theta,$$

where  $\mathbb{Z}[i]\left[\frac{1+\sqrt{5}}{2}\right] = \{a + b\sqrt{5} \mid a, b \in \mathbb{Z}[i]\}$ .

## Crossed product algebras

- Codes for *4 antennas*: take  $L/K$ , with

$$L = K(\sqrt{d}, \sqrt{d'}), \text{ Gal}(L/K) = \{1, \sigma, \tau, \sigma\tau\}.$$

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- A *crossed product algebra*  $\mathcal{A} = (a, b, u, L/K)$  over  $L/K$ :

$$\mathcal{A} = L \oplus eL \oplus fL \oplus efL$$

with

$$e^2 = a, \quad f^2 = b, \quad fe = efu, \quad \lambda e = e\sigma(\lambda),$$

$$\lambda f = f\tau(\lambda) \text{ for all } \lambda \in L,$$

for some elements  $a, b, u \in L^\times$  satisfying

$$\sigma(a) = a, \tau(b) = b, u\sigma(u) = \frac{a}{\tau(a)}, u\tau(u) = \frac{\sigma(b)}{b}.$$

## Codewords from crossed product algebras

- Let  $x = x_1 + ex_\sigma + fx_\tau + ef x_{\sigma\tau} \in \mathcal{A}$ . Its left *multiplication matrix*  $X$  is given by

$$\begin{pmatrix} x_1 & a\sigma(x_\sigma) & b\tau(x_\tau) & ab\tau(u)\sigma\tau(x_{\sigma\tau}) \\ x_\sigma & \sigma(x_1) & b\tau(x_{\sigma\tau}) & b\tau(u)\sigma\tau(x_\tau) \\ x_\tau & \tau(a)u\sigma(x_{\sigma\tau}) & \tau(x_1) & \tau(a)\sigma\tau(x_\sigma) \\ x_{\sigma\tau} & u\sigma(x_\tau) & \tau(x_\sigma) & \sigma\tau(x_1) \end{pmatrix}.$$

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- Such codewords are *fully-diverse* if  $\mathcal{A}$  is a division algebras.

## A criterion for full-diversity

**Theorem.** Let  $K$  be a number field, and let  $\mathcal{A} = (a, b, u, L/K)$ . Then the following conditions are equivalent:

1.  $\mathcal{A}$  is a division algebra,
2. the quaternion algebra  $(d, N_{K(\sqrt{d'})/K}(b))$  is not split,
3. the quaternion algebra  $(d', N_{K(\sqrt{d})/K}(a))$  is not split.



## Encoding

- Let  $\{\omega_1, \omega_2, \omega_3, \omega_4\}$  be a  $\mathbb{Q}(i)$ -basis of  $L$ ,  $G$  be the matrix of the embeddings of the basis,  $\mathbf{x} = (x_1, x_2, x_3, x_4)$  be 4 information symbols,  $x = x_1\omega_1 + x_2\omega_2 + x_3\omega_3 + x_4\omega_4 \in L$ .

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- We encode 16 information symbols  $G\mathbf{x}_1, G\mathbf{x}_\sigma, G\mathbf{x}_\tau, G\mathbf{x}_{\sigma\tau}$  with

$$G\mathbf{x} = (x, \sigma(x), \tau(x), \sigma\tau(x))^T.$$

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- We encode 16 information symbols  $G\mathbf{x}_1, G\mathbf{x}_\sigma, G\mathbf{x}_\tau, G\mathbf{x}_{\sigma\tau}$  with

$$G\mathbf{x} = (x, \sigma(x), \tau(x), \sigma\tau(x))^T.$$

- Define  $\Gamma_1 = \mathbf{I}_4$ , and  $\Gamma_j, j = 2, 3, 4$  resp. as

$$\begin{pmatrix} 0 & a & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & \tau(a) \\ 0 & 0 & 1 & 0 \end{pmatrix}, \begin{pmatrix} 0 & 0 & b & 0 \\ 0 & 0 & 0 & b\sigma(u) \\ 1 & 0 & 0 & 0 \\ 0 & \sigma\tau(u) & 0 & 0 \end{pmatrix}, \begin{pmatrix} 0 & 0 & 0 & ab\sigma(u) \\ 0 & 0 & b & 0 \\ 0 & \tau(a)\tau(u) & 0 & 0 \\ 1 & 0 & 0 & 0 \end{pmatrix}.$$

The codeword  $X$  is encoded as follows:

$$X = \Gamma_1 \text{diag}(G\mathbf{x}_1) + \Gamma_2 \text{diag}(G\mathbf{x}_\sigma) + \Gamma_3 \text{diag}(G\mathbf{x}_\tau) + \Gamma_4 \text{diag}(G\mathbf{x}_{\sigma\tau}).$$

## Example of code

- Consider the algebra on  $\mathbb{Q}(i)(\sqrt{2}, \sqrt{5})/\mathbb{Q}(i)$ .
- We take

$$a = \zeta_8, \quad b = \sqrt{\frac{1+2i}{1-2i}}, \quad u = i.$$

Thus the encoding matrices  $\Gamma_i$ ,  $i = 2, 3, 4$  are *unitary*.

## Example of code

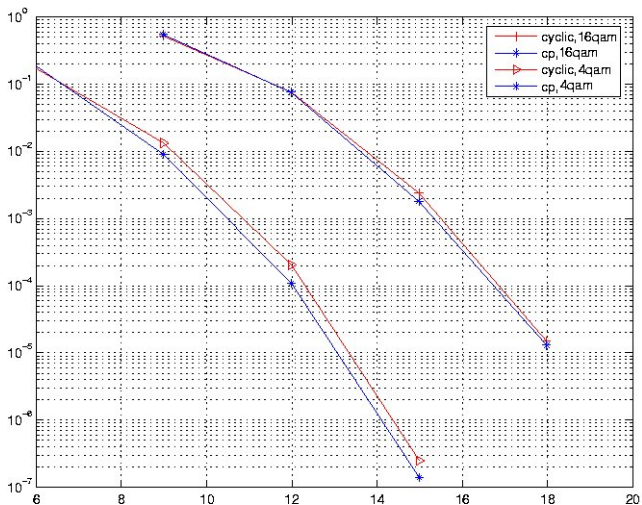
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Thus the encoding matrices  $\Gamma_i$ ,  $i = 2, 3, 4$  are *unitary*.

- We obtain a matrix  $G$  *unitary* by restricting to an ideal of  $L$ .
- This is a division algebra.

# Comparison with previous codes



# An Order View Point

- Replace copies of  $\mathcal{O}_K$  by a maximal order with minimized discriminant.

[ R. Vehkalahti, C. Hollanti, J. Lahtonen, K. Ranto, *On the densest MIMO lattices from cyclic division algebras.* ]

# Summary

To obtain fully diverse space-time codes from division algebras:

1. For  $n$  antennas, consider a cyclic extension of  $\mathbb{Q}(i)$ , or for  $n = 4$ , a biquadratic extension of  $\mathbb{Q}(i)$ . Construct a cyclic/crossed product division algebra.
2. Restrict coefficients to the ring of integers (minimum determinant).
3. Add lattices on each “layer”.



## Division Algebras

Cyclic Algebras

Crossed Product Algebras

## Quotients of Space-Time Codes

$2 \times 2$  Space-Time Coded Modulation

# $2 \times 2$ MIMO Slow Fading Channel

$$\underbrace{\mathbf{Y}}_{2 \times 2L} = \underbrace{\mathbf{H}}_{2 \times 2} \mathbf{X} + \underbrace{\mathbf{Z}}_{2 \times 2L}$$

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- $2L =$  frame length.
- $\mathbf{X} = [X_1, \dots, X_L] \in \mathbb{C}^{2 \times 2L}$ .

## Code Design Criteria

Design

$$\mathbf{X} = [X_1, \dots, X_L] \in \mathbb{C}^{2 \times 2L}$$

such that

1.  $X_i$  are fully diverse,  $i = 1, \dots, L$ .
2. the minimum determinant

$$\begin{aligned} \Delta_{min} &= \min_{\mathbf{0} \neq \mathbf{X}} \det(\mathbf{X}\mathbf{X}^*) \\ &= \min_{\mathbf{0} \neq \mathbf{X}} \det\left(\sum_{i=1}^L X_i X_i^*\right) \\ &\geq \min_{\mathbf{0} \neq \mathbf{X}} \left(\sum_{i=1}^L |\det(X_i)|\right)^2 \end{aligned}$$

is maximized.

# Concatenated codes

1. Choose  $X_i$ ,  $i = 1, \dots, L$  *independently*.

# Concatenated codes

1. Choose  $X_i$ ,  $i = 1, \dots, L$  *independently*.
2. Use a *concatenated code*:
  - *inner code* for diversity
  - *outer code* for coding gain

[ L. Luzzi et al., *Golden Space-Time Block Coded Modulation* ]

## One example: the Golden Code $\mathcal{G}$

- The *inner code*:

$$X = \frac{1}{\sqrt{5}} \begin{pmatrix} \alpha(a + b\theta) & \alpha(c + d\theta) \\ i\sigma(\alpha)(c + d\sigma(\theta)) & \sigma(\alpha)(a + b\sigma(\theta)) \end{pmatrix} \in \mathcal{G}$$

- $a, b, c, d \in \mathbb{Z}[i]$ ,  $\theta = \frac{1+\sqrt{5}}{2}$ ,  $\sigma(\theta) = \frac{1-\sqrt{5}}{2}$ ,  $\alpha = 1 + i - i\theta$  and  $\sigma(\alpha) = 1 + i - i\sigma(\theta)$ .

## Coset codes

- We have  $\mathcal{G} = \alpha(\mathbb{Z}[i, \theta] \oplus e\mathbb{Z}[i, \theta])$ ,  $e^2 = i$  and (more later)

$$\mathcal{G}/(1+i)\mathcal{G} \simeq \mathcal{M}_2(\mathbb{F}_2).$$



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- Construct a code on  $\mathcal{M}_2(\mathbb{F}_2)$  and lift it (*outer code*).
- For a coset code (Luzzi et al.)

$$\Delta_{min} \geq \min_{\mathbf{0} \neq \mathbf{X}} \left( \sum_{i=1}^L |\det(X_i)| \right)^2 \geq \min (|1+i|^4 \delta, d_{min}^2 \delta),$$

$\delta =$  minimum determinant of  $\mathcal{G}$ ,  $d_{min} =$  minimum distance.

## Linking $\mathcal{M}_2(\mathbb{F}_2)$ and $\mathbb{F}_4$

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where  $j^2 = 1$  and  $j\omega = \omega^2j$ , given by

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- This means:

$$\phi : (a, b) \in \mathbb{F}_4 \times \mathbb{F}_4 \mapsto M_{a,b} \in \mathcal{M}_2(\mathbb{F}_2).$$

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- $\phi : (a, b) \in \mathbb{F}_4 \times \mathbb{F}_4 \mapsto M_{a,b} \in \mathcal{M}_2(\mathbb{F}_2)$  maps

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- Define a weight on the matrices

$$w(M_{a,b}) = \begin{cases} 0 & M_{a,b} = 0 \\ 1 & M_{a,b} \text{ invertible} \\ 2 & 0 \neq M_{a,b} \text{ non-invertible} \end{cases} .$$

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- $\phi$  is an isometry:

$$w(M_{a,b}) = w(\phi((a, b))) = w_H((a, b))$$

where  $w_H$ =Hamming weight.



## Back to the outer code design

- For a coset code

$$\Delta_{min} \geq \min \left( 4\delta, \frac{w_{min}^2}{2} \delta \right),$$

$\delta$  = minimum determinant of  $\mathcal{G}$ ,  $w_{min}$  = minimum weight on code over  $\mathbb{F}_4$ .

## Example

- Take the  $[6,3,4]$  hexacode over  $\mathbb{F}_4$ , with

$$y = (y_1, y_2, y_3, y_1 + \omega(y_2 + y_3), y_2 + \omega(y_1 + y_3), y_3 + \omega(y_1 + y_2)).$$

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- Compute  $\phi((y_1, y_2))$ .

$$\begin{aligned} (y_1, y_2) &\mapsto y_1 + y_2j = (y_{11} + y_{12}\omega) + (y_{21} + y_{22}\omega)j \\ &\mapsto \begin{pmatrix} y_{11} & y_{12} \\ y_{12} & y_{11} + y_{12} \end{pmatrix} + \begin{pmatrix} y_{21} & y_{22} \\ y_{22} & y_{21} + y_{22} \end{pmatrix} \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \\ &= Y_1 \end{aligned}$$

- 

$$\phi(y) = (Y_1, Y_2, Y_3),$$

with minimum weight  $w_{min} = 4$ .

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- For coding for *MIMO slow fading channels*, joint design of an *inner and outer* code.
- The outer code is a *coset code*, which addresses the problem of *codes over matrices*.
- Connection between *codes over matrices* and *codes over finite fields*.

Thank you for your attention!